

DC-Distributed Power System Modeling and Hardware-in-the-Loop (HIL) Evaluation of Fuel Cell-Powered Marine Vessel

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Abstract—Environmentally driven regulations are significantly affecting shipping in recent years, where the shipbuilding industry is required to comply with upcoming restrictions concerning polluting emissions. The all-electric ship (AES) is one of the most promising technologies for complying with the increasingly strict environmental regulations, improving fuel efficiency, and enhancing system dynamic performance. In this study, the dc-distributed power grid of an AES integrated with fuel cells and batteries has been configured using extensive electrification technology, where the system-level shipboard power plant has been modeled with the average modeling method. The model not only incorporates the hybrid power source integration but also the primary and secondary power management as a whole. In addition, a hardware-in-the-loop (HIL) has been set up to replicate the real-time system behavior, which is essential for the verification of any optimal power management control algorithms to be developed in future work. Finally, both the mathematical and real-time models are validated against the full-scale hybrid shipboard power system.

Index Terms—All-electric ship, fuel cell system (AES), hardware-in-the-loop (HIL), hybrid shipboard power microgrid, renewable energy sources.

I. INTRODUCTION

MARITIME transport has played a dominant role in the global trade system for centuries, and it is expected that

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international maritime trade will expand at continuous annual growth for years to come [1]. Recently, environmental sustainability has become a significant policy concern in global maritime transport [2]. With this concern, the shipbuilding industries are compelled to abide by rules on polluting emissions imposed by International Maritime Organization (IMO) and national legislation [3]. To achieve high energy efficiency and low emission, it is believed that the all-electric ship (AES) will be one of the most promising technologies to comply with the environmental regulations [4]. In particular, vessel types such as tug boats, river pushers, and ferries will be the early adopters as they usually operate near coasts or in inner rivers, which belong to emission control areas (ECAs). Other vessel types such as container vessels may need more time before the AES design can be adopted [5].

With the development of energy storage technologies, new energy sources can be integrated into the shipboard power network. Batteries, as the most mature energy storage system (ESS) device, have been widely utilized in different applications in marine vessels. However, due to energy density limitation, battery systems only function as a main power supply for small-scale vessels in short distance shipping segment or as an auxiliary power source onboard of large vessels to enhance the dynamic performance or short-term zero-emission operation. In order for clean energy sources to be a feasible solution for larger ocean-going ship types, such as cruise ships and RoRo/RoPax (the vessels built for freight vehicle transport with passenger accommodation), fuel cell systems have been introduced. It is expected that fuel cell systems would replace diesel engines as the main power source in the future.

The hybrid shipboard power system integrated with different types of power sources brings in a lot of benefits, but it also increases the complexity of the system structure and power management control. To get deep understanding of the hybrid power system and improve the current control algorithms, a comprehensive system-level hybrid power plant model is needed as a foundation. The expected system model should be able to perform the actual shipboard power plant responses under different operating modes and loading profiles.

A few power system configurations and modeling methodologies have been published in the recent literature. In [6] and [7], land-based hybrid dc microgrid configurations have

72 been introduced. The authors in [8] have proposed a diesel-
 73 generator-based marine power network equipped with a battery
 74 ESS, while a fuel cell-based marine system model has been
 75 developed in [9], where the authors are mainly focusing on
 76 the fuel cell chemical reaction and fuel processing. The above
 77 references focus only on certain parts of the power grid rather
 78 than the observation of the dynamics and performance of the
 79 complete system.

80 It is worth noting that shipboard power systems have their
 81 own special configurations and requirements [10]. The main
 82 difference with land-based power system is that shipboard power
 83 system is an isolated microgrid with short distances from gen-
 84 erated power to the electric propulsion load [11]. The average
 85 short-circuit levels and forces are quite high so that it must be
 86 dealt with some special manners to protect the power equipment.
 87 During cruising, shipboard power system is impacted a lot by the
 88 environment, such as marine weather and unexpected thrusters
 89 load variations, therefore the system stability and reliability is
 90 crucial for marine vessels. Extra efforts in system reconfigurabil-
 91 ity and redundancy are required to prevent single point failure
 92 according to the marine classification rules. Compared to EV
 93 power system, the power scale of the shipboard power network
 94 is much bigger and complicated. Due to its particularity of the
 95 shipboard power system [12], there is the urge to build its own
 96 power plant model rather than reusing other existing models.

97 In addition, system-level power plant model is not only to inte-
 98 grate different power sources but also to implement primary and
 99 secondary level power management to coordinate power load
 100 sharing, frequency, and voltage control. A previous work by the
 101 authors [13] modeled an ac-distributed shipboard hybrid power
 102 system. However, in the past decade, dc-based distribution sys-
 103 tems have gained popularity due to their advantages in many
 104 aspects [14]. A system-level dc-based shipboard power model
 105 is necessary to simulate the vessel operating performances. The
 106 comprehensive and reliable power plant model is the foundation
 107 and precondition for system studies and further research.

108 Compared to the existing studies on hybrid power network
 109 modeling problems, the main contributions of this article are
 110 highlighted as follows.

- 111 1) A dc-distributed system-level shipboard power system
 112 model is built, integrating the mathematical model of differ-
 113 ent power sources, as well as the primary and secondary
 114 power management control. All the power sources are res-
 115 izable to configure the new shipboard systems according
 116 to the varied ship design specifications.
- 117 2) A hardware-in-the-loop (HIL) platform is set up in lab
 118 environment as well. HIL helps to perform the system
 119 real-time states. It is a useful tool for prototyping the power
 120 and energy management system (PEMS) controller and
 121 communication interface validation without involving any
 122 hardware devices setup.
- 123 3) Both the mathematical and real-time HIL model have
 124 been validated against the real hardware full-scale hybrid
 125 shipboard power system.

126 The rest of this article is organized as follows: The hybrid
 127 shipboard power system configuration is discussed in Section II.
 128 A dc-based power distribution model and the simulation results

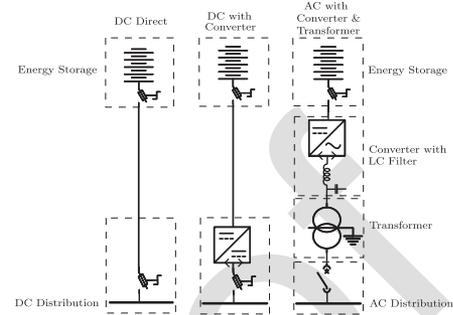


Fig. 1. ESS integration scheme for both dc and ac distribution system.

are presented in Section III. Section IV evaluates the system
 plant model HIL setup. And the validation tests are performed
 in Section V. Finally, Section VI concludes this article.

II. SHIPBOARD POWER SYSTEM CONFIGURATION

In this section, the functions of several integration schemes
 will be introduced in detail. The advantages and the system
 configuration of dc-distributed shipboard power system will also
 be discussed.

A. Advantages of DC-Based Shipboard Power System

Currently, the majority of commercial marine vessels have a
 diesel or gas engine powered plant with ac distribution, while
 the dc-distributed shipboard system has drawn much attention
 over the past decade. In both ac and dc-distributed systems, ESS
 integration is a trend for different functionalities [15].

Energy storage, such as batteries will help the slow power
 devices level out load variations from the thrusters and other
 heavy consumers. Besides, the clean energy sources are able to
 provide the main power supply of the ship operation in order
 to achieve zero-emission operation. There are several methods
 to integrate the ESS into the shipboard network [16], which is
 shown in Fig. 1.

According to this integration scheme, it can be seen that the
 ESS is much easier to connect to a dc-distributed network than an
 ac system. The storage devices can be directly connected to the
 dc network. This direct-on-line (DOL) ESS gives an efficiency
 increase compared to the ESS with dc/dc converter, but at the
 cost of less control features to be achieved. On the other hand,
 ESS is also possible to be extended to an ac-distributed system
 through a dc/ac converter with LCL filter and transformer [17].
 The LCL filter is used herein to improve the power quality and
 reduce the total harmonic distortion (THD) level of the main
 power network. While the shielded transformer is utilized to
 solve the electromagnetic interference (EMI) issue, it also blocks
 the circulating current within the power electronic converters
 such as IGBT or MOSFET. In addition, the dc/ac converters also
 need to monitor the amplitude, frequency, and phase shift of the
 power network voltage on ac side to achieve the synchronization
 operation [13]. It is clear that the ac-based power configuration
 is quite complex not only in system configuration but also in
 power management control [14]. Furthermore, it also increases

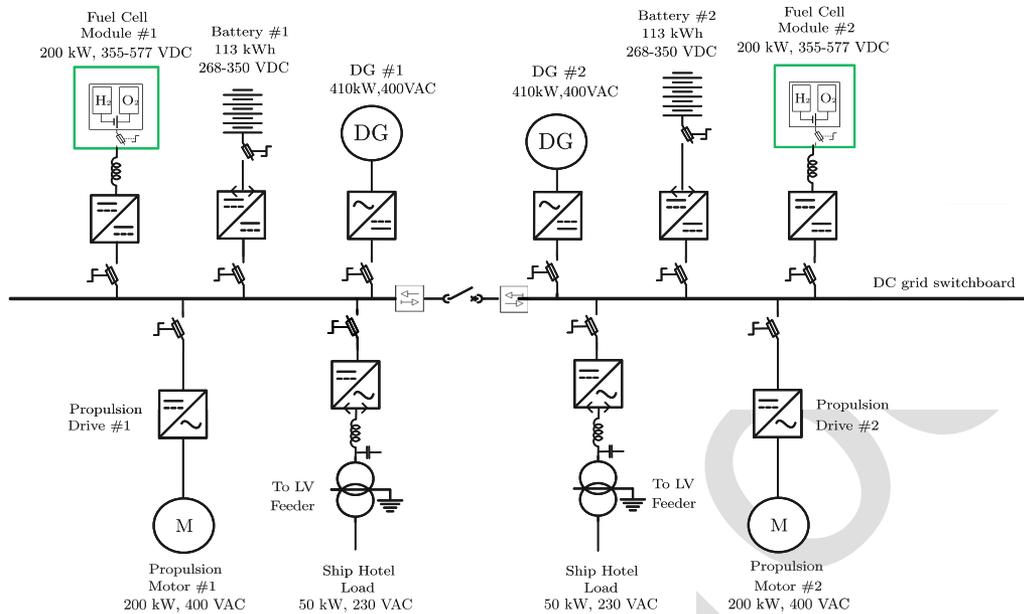


Fig. 2. Typical dc-based shipboard power system single line diagram.

the dimensions of the equipment, where the space onboard a ship is limited.

Safety is another important consideration for implementing the dc-based distribution. A dc-configured system is inherently simpler than an ac system, which means that it is easier to predict fault scenarios and devise adequate protection against them. The generators can be connected to the dc distribution system in a short time because it is not necessary for synchronization. It has considerable potential to improve the stability, efficiency, and performance of future dc-based ship power systems [18].

In addition to the benefits described above, there are numerous other benefits with dc distribution. Some of these are summarized as follows.

- 1) *Voltage Distortion*: Harmonic distortion is inherent in ac systems with frequency converters, while being less of a concern with dc distribution.
- 2) *Electrical Efficiency*: In the process of going from ac-to-dc distribution, the system efficiency is improved by 0.5–1% [19].
- 3) *Ease of Control*: It is not required to control the frequency and reactive power of the network. The control system is simplified and only needs to overlook the dc voltage.

Overall, the integration of a hybrid ac-distributed shipboard power is complex and costly [14]. Multiple layers of the power conversion and transmission increase the system losses and probability of equipment failure. It is more suitable for some retrofit projects, which are planned for extending a single ESS from an existing ac distribution network. Due to the advantages of dc-distribution power system, this study focuses on dc-based shipboard power plant modeling.

B. Hybrid Shipboard Power System Configuration

For a hybrid dc-distributed shipboard power system, the energy storage devices can be connected to the system. A generator set driven by a combustion diesel engine is possible to be integrated into the dc power network. Usually, synchronous generators are used and rectified (ac to dc converter) before the power is transmitted to the dc bus. Propulsion and other ship service loads are regulated by dc/ac inverters. The typical configuration of an all-electric dc hybrid ship power and propulsion system is shown in Fig. 2. In this study, two fuel cell modules are function as the main power supply to achieve zero emission operation. Two battery banks handle the ship load variations. And two diesel gen sets (DG) are the backup power in case the fuel cell devices or the hydrogen fuel are not available under emergency condition.

The main switchboards are usually split into two sections or more to obtain the redundancy requirements of the vessel according to the marine class rules and regulations. This kind of structure is not only an intrinsic advantage of dc distribution but it also has the ability to supply power to the load continuously even under certain fault conditions [16], such as short-circuit. During normal operation, the dc-grid switchboard are usually connected (close bus tie), which gives the best flexibility in the configuration of the power generation. The optimal number of the power devices can be connected to the power network according to the load transients to achieve certain control criteria.

III. FUEL CELL-FED DC DISTRIBUTION SYSTEM MODELING

In this section, a system-level dc-based hybrid shipboard power plant will be modeled. The power network includes the fuel cells, batteries, and DGs as the energy sources. The essential components of the dc-grid power system are the power

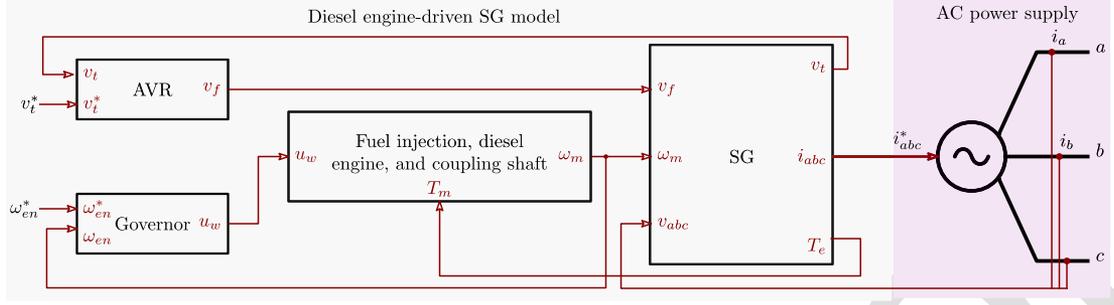


Fig. 3. General block diagram of a diesel engine-driven synchronous generator.

electronics devices. An ESS dc/dc converter and a 6-pulse ac/dc rectifier for DG integration will be modeled. The propulsion load and ship hotel loads of the vessel will be represented with controllable current sources. Power management, including load power sharing, dc voltage coordination, and power regulation will be achieved as well.

A. Energy Sources Modeling

1) *Diesel Gen-Set Model*: Fig. 3 shows the general configuration of a diesel engine-driven synchronous generator system. The system consists of four main sections: Fuel injection and diesel engine, governor, synchronous generator (SG), and automatic voltage regulator (AVR) [20]. The mathematical models of each part of diesel generator are introduced as follows.

a) *Fuel injection and diesel engine* can be presented by a time delay and the coupling shaft model, as in [21] and [22]. The simplified model is given by

$$\frac{T_m(s)}{u_w(s)} = \frac{K_{en} \cdot e^{-t_d \cdot s}}{t_e s + 1} \quad (1)$$

$$J_{eq} \cdot \frac{d\omega_m}{dt} = T_m - T_e - k_{f_{eq}} \cdot \omega_m \quad (2)$$

where ω_m is the mechanical speed, u_w is the control signal from the speed governor, T_m is the mechanical torque developed by the engine, J_{eq} is the equivalent inertia of the entire system, $k_{f_{eq}}$ is the equivalent friction coefficient, K_{en} is the engine gain, t_d is a delay representing the time elapsed from the fuel injection until the torque is developed at the engine shaft, and t_e is the time constant of the fuel injection.

b) *Governor* is responsible for regulating the engine speed to maintain it within the allowable range during the different load conditions, resulting in constant output frequency or within the desired limits. This governor also has the structure of a PI controller with the droop function implemented. This model can be given by

$$u_w = (\omega_{en}^* - \omega_{en} - k_{dr, \text{freq}} \cdot u_w) \left(K_{P_\omega} + \frac{K_{I_\omega}}{s} \right) \quad (3)$$

where K_{P_ω} and K_{I_ω} are the proportional and integral gains of the governor PI controller, and ω_{en} is the engine nominal speed and $k_{dr, \text{freq}}$ is the speed droop

gain that equals $(m_{dr} \cdot \omega_{en})$, and m_{dr} is the static droop slope.

c) *Synchronous generator* is considered without damper windings for the sake of simplifying the model. And its d - q rotating reference frame is given by

$$V_d = -r_s \cdot i_d + L_q \cdot \omega_e \cdot i_q + L_d \cdot \frac{di_d}{dt} + M_{sf} \cdot \frac{di_f}{dt} \quad (4)$$

$$V_q = -r_s \cdot i_q - L_d \cdot \omega_e \cdot i_d + L_q \cdot \frac{di_q}{dt} + M_{sf} \cdot \omega_e \cdot i_f \quad (5)$$

$$V_f = r_f \cdot i_f + L_f \cdot \frac{di_f}{dt} - M_{sf} \cdot \frac{di_d}{dt} \quad (6)$$

$$T_e = (L_d - L_q) \cdot i_d \cdot i_q + M_{sf} \cdot i_q \cdot i_f \quad (7)$$

where V_d, V_q, i_d , and i_q are the output voltages and currents in the d - q rotating reference frame, respectively, V_f and i_f are the field excitation voltage and current, respectively, r_s and r_f are the stator winding and field winding internal resistances, respectively, L_d and L_q are the stator inductance in the d -axis and q -axis, respectively, L_f is the inductance of the field, M_{sf} is the mutual inductance between the field winding and the d -axis stator winding, and T_e is the electromagnetic torque. It is worth noting that, the following assumptions have been considered: The stator windings are symmetrical, a uniform sinusoidal distribution along the air gap, the permanence of the magnetic paths on the rotor is independent of the rotor positions, and the saturation and the hysteresis effects are neglected.

d) *Automatic voltage regulator* has the responsibility of controlling the terminal voltage of the synchronous generator under different load conditions. This AVR is modeled as a first-order system, representing a power converter controlled by a PI controller [20], that is given by

$$V_f = \frac{k_{conv}}{t_{conv} s + 1} \cdot (U_v - k_{dr, v} V_f) \quad (8)$$

$$U_v = \left(K_{P_v} + \frac{K_{I_v}}{s} \right) \cdot (V_t^* - V_t) \quad (9)$$

where K_{P_v} and K_{I_v} are the proportional and integral gains of the AVR PI controller, k_{conv} and t_{conv} are the

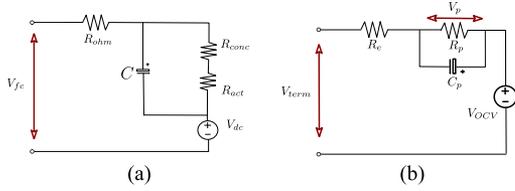


Fig. 4. ESS power source modeling. (a) Proton-exchange membrane (PEM) fuel cell and its equivalent circuit. (b) Lithium-ion battery and its equivalent circuit.

converter gain and time constant, and V_t^* and V_t are the reference and measured rms output line voltage. $k_{dr,v}$ is the dc voltage droop rate for dc-distributed power system.

- 2) *Fuel Cell Model*: There are a few different types of fuel cell technologies. The proton-exchange membrane (PEM) fuel cell is the most mature technology over other types of fuel cell. With its relatively low cost and high energy efficiency features, PEMFC is the most widely used in marine applications [23]. Fig. 4(a) shows a single PEM fuel cell model and its equivalent circuit [24]. The fuel cell losses are mainly divided into three categories: Activation loss R_{act} , concentration loss R_{conc} , and ohmic loss R_{ohm} [25]. The voltage of a PEMFC elementary cell can be written as follows [26]:

$$V_{cell} = E_{nemst} - V_{act} - V_{conc} - V_{ohm} \quad (10)$$

$$V_{fc} = N_{fc} \times V_{cell} \quad (11)$$

where V_{cell} is the voltage of a PEMFC elementary cell, E_{nemst} is the equilibrium voltage, V_{act} is the activation overpotential, V_{conc} is the concentration overpotential, V_{ohm} is the ohmic overpotential, V_{fc} is the voltage of PEMFC stack, and N_{fc} is the number of cells in series.

In this study, the PEMFC is configured with Matlab/Simulink Simscape library.

- 3) *Battery Model*: To balance the cost, energy density, power density, safety performance, and life span, Li-ion battery is the most applied battery type in marine ESS applications [27]. Simple models based on the equivalent circuit of a resistor in series with a parallel RC circuit (polarization circuit) [28] have been used to determine the dynamic characteristics of the battery, as shown in Fig. 4(b). Open-circuit voltage (OCV) is defined differently for discharging and charging at given state-of-charge (SoC) due to the hysteresis effect [29]. The total voltage drop from OCV is then modeled here based on the first-order equivalent circuit to account for the internal resistance and polarization (OCV relaxation).

$$V_{term} = V_{ocv} - IR_e - V_p \quad (12)$$

where V_{term} is the terminal voltage of the cell, V_{ocv} is the cell OCV, I is the current rate of the cell (positive for discharging and negative for charging), R_e is the electrical resistance (responsible for instantaneous voltage drop together with C_p , after the application of load) and V_p is the voltage drop across the polarization circuit.

At any instant, V_p across the parallel RC circuit is represented by the voltage drop across R_p or C_p . So, the equation can be arranged by

$$\frac{V_p(t)}{R_p} + C_p \frac{dV_p(t)}{dt} = I(t) \quad (13)$$

where I_{Rp} is the current across polarization resistance, and I_{Cp} is the current across polarization capacitance. The sum of I_{Rp} and I_{Cp} gives the total circuit current I . Then perform Laplace transform by

$$V_p(s) = I(s) \frac{R_p}{1 + sR_pC_p} \quad (14)$$

so that the steady-state response of terminal voltage $V_{term s-s}$ to a given current by

$$V_{term s-s} = V_{ocv} - I(R_e + R_p) = V_{ocv} - IR_{ti} \quad (15)$$

where $R_{ti} = R_e + R_p$ is the total internal resistance of a cell.

SoC is another critical parameter for the battery system. SoC is intrinsically divided into two types: Instantaneous SoC (SoC_{inst}) and nominal SoC (SoC_{nom}), both based on Ah counting. Nominal SoC is derived from nominal capacity based on standard discharging conditions as suggested by the manufacturer, while instantaneous SoC is based on the available instantaneous discharge capacity given by

$$SoC_{nom}(t_i) = SoC_{nom}(t_{i-1}) - \frac{I_{step}(t_{i-1})t_{step}}{C_{nom}} \quad (16)$$

$$SoC_{inst}(t_i) = \frac{(1 - SoC_{nom}(t_{i-1}))C_{nom}}{C_{step\ avail}(t_{i-1})} - \frac{I_{step}(t_{i-1})t_{step}}{C_{step\ avail}(t_{i-1})} \quad (17)$$

where, I_{step} is the current at a given time step, t_{step} is the span of time under consideration (sampling time of BMS/controller), and $C_{step\ avail}$ is the capacity available for I_{step} . Instantaneous SoC is useful in indicating the capacity available at a given time step and operating conditions, especially higher discharge rates.

In this study, the Matlab/Simulink Simscape battery stack model is utilized.

B. Power Electronic Converter Modeling

A dc/dc converter (for fuel cell and battery control), and an ac/dc rectifier (for diesel generator control) are modeled in this section.

- 1) *DC/DC Converter Model*: Both of the fuel cells and the battery systems are powered by dc/dc converter. The dc/dc converter regulates the output voltage and power flow from the storage device [8]. Different from the battery bidirectional control [31], fuel cells only require a single direction power flow to provide the power to dc link. It can be represented as the following equation:

$$\frac{V_{dc}}{V_{ESS}} = \frac{1}{1 - D} \quad (18)$$

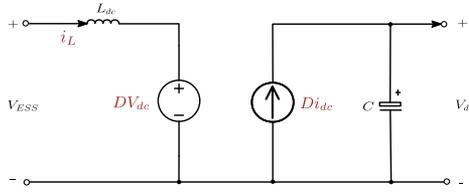


Fig. 5. DC/DC converter average model.

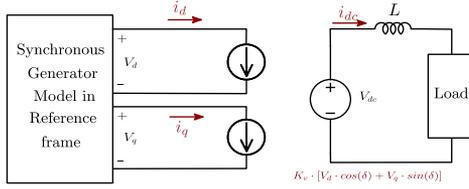


Fig. 6. Average model of 6-pulse ac/dc rectifier, where V_d , V_q , I_d , I_q are generator's output voltage and current in $dq - axes$, δ is the angle between V_d and V_q , coefficient K_v is the ratio between the magnitude of rectifier output voltage V_{dc} and its input voltage V_d , V_q , and δ .

where D is the duty cycle of dc/dc converter.

There are two methods to model the power electronic converters, i.e., electromagnetic transient (EMT) model and average model. EMT model is straightforward and can represent the modulation process of the power electronic device. However, the simulation speed will be slow down due to the usage of high-frequency carriers during pulsewidth modulation (PWM). EMT simulation method is more suitable for system stability analysis studies, but not suitable to evaluate system performance throughout the operational profile which can last in the order minutes or even hours. For that reason, the average model is selected to configure the shipboard power plant model here. The dc/dc converter average model is shown in Fig. 5.

- 2) *AC/DC Rectifier Model*: The DGs are connected into dc bus via a 6-pulse diode rectifier. The average model of a diode bridge loaded synchronous generator was used here to simplify the calculation of the average values of generator $d-q$ variables to the dc voltage and current at the bridge output [32], as shown in Fig. 6.

The voltage and current at the dc terminal can be defined as proportional to the amplitudes of the first harmonics of generator phase voltage and current, respectively [33]

$$V_{dc} = K_v \sqrt{V_d^2 + V_q^2} \quad (19)$$

$$i_{dc} = K_i \sqrt{i_d^2 + i_q^2} \quad (20)$$

where V_{dc} is dc voltage output from the rectifier. V_d , V_q , i_d , and i_q are the voltage and current of generator in dq -axes. K_v is the generator voltage constant, and K_i is the generator current constant.

C. System Control Philosophy

For existing hybrid shipboard power management is usually arranged in line with the hierarchical control framework. The basic structure is shown in Fig. 7.

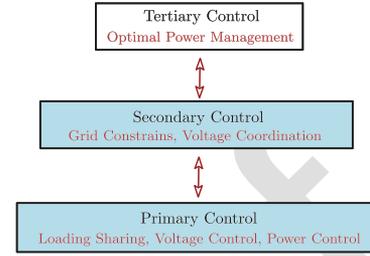


Fig. 7. Hierarchical control framework for shipboard power management, where the primary and secondary control have been applied.

In hierarchical control, the primary control level is to handle the load sharing among the hybrid power sources. The secondary control level includes the grid voltage coordination and to ensure bus signals at their operating ranges. The tertiary control level is used to achieve optimal operation with intentional objectives [34]. In this study, the primary and secondary level power management control have been achieved, while the tertiary control will be considered in future's research work.

The basic load sharing mechanism in dc-distributed power network is dc voltage droop control. Basically, with the voltage droop control, all the power sources share the load based on the dc bus voltage in the similar way as with frequency droop control in ac-distributed system. The load sharing function is realized by adjusting the voltage droop rate. In Fig. 8, it shows the load sharing results of a dc power network with three different droop rate setting. A steeper droop angle will cause one consumer to take less load than one with a more flat droop angle given the same voltage set point.

DG system is integrated into the dc network via 6-pulse diode rectifier, and the output voltage is controlled by AVR. The dc voltage set point is defined as the nominal dc voltage of the rectifier. Droop increases with increasing generator speed. Usually, the dc voltage droop rate is not larger than 5% [35].

For the energy storage power sources, such as fuel cells and batteries, the dc voltage droop control is implemented in the power electronic dc/dc converters. The block diagram of the voltage control loop is as shown in Fig. 9. It is a dual control loop, with the current control as the inner loop with the voltage control as the outer loop. $K_{dr,v}$ is the voltage droop rate for load sharing with other power sources.

The dc voltage droop control is simple and reliable. But due to the complexity of hybrid dc-grid system configuration, it is difficult to fine-tune the whole power system with one single parameter to maintain the performance of each power source, and ensure the system stability at the same time. It is also not possible to achieve strategy loading or even optimized operation. Therefore, the power control of the ESS devices is expected. The dc/dc converter power control loop with PI controllers is shown in Fig. 10. It is a single current control loop, where the current reference is given by the quotient of reference power and dc voltage.

In this study, dc voltage droop combined with power control is implemented as primary power management scheme. The system voltage coordination rules and constrains have been set in power electronic converter models as the secondary level of

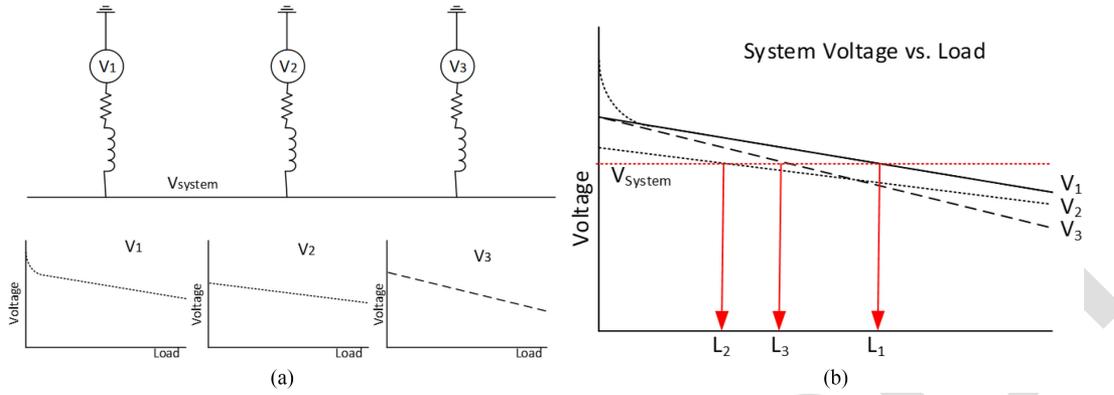


Fig. 8. Load sharing with 3 different droop curves. (a) Integrated dc-distributed system with three power sources and their droop rates. (b) System voltage and load sharing scheme.

TABLE I
PARAMETERS FOR THE DC-BASED DEMO VESSEL

Diesel-generator		Fuel cell [30]		Batteries	
Rated power (P_{gen})	410 kW	Rated power (P_{fc})	2*100 kW	Capacity (E_{batt})	113 kWh
Nominal voltage (V_{abc})	400 VAC	Min. power (P_{fc_min})	20 kW	Max. charge voltage (V_{batt_max})	350VDC
Nominal current (I_{abc})	592 A	Operating voltage (V_{fc})	355-577 VDC	Nominal voltage (V_{batt})	300 VDC
Frequency (F_n)	50 Hz	Rated current (I_{fc})	2*257 A	Min. voltage (V_{batt_min})	268 VDC
Speed (Spd)	1500 rpm	Max. Temperature (T_{max})	70°C	Max. discharge current (I_{batt_disch})	250 A
Inertia (J_{ge})	10 kgm ²	Fuel Type	Hydrogen	Max. charge current (I_{batt_ch})	200 A
Power factor (pf)	0.9	H ₂ pressure	3.5-5 Barg	Operating state of charge range (SoC)	20%-80%

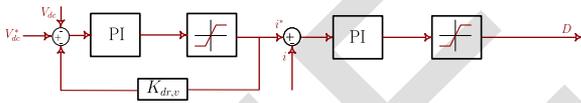


Fig. 9. DC/DC converter control scheme I - Voltage control loop.

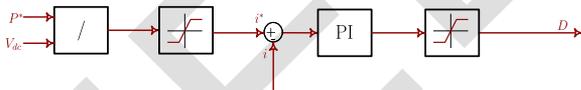


Fig. 10. DC/DC converter control scheme II - Power control loop.

power management. The ESS device output power or voltage shall be regulated according to the reference command if the corresponding control scheme is selected. Two different operation modes will be studied in this article: Zero-emission mode and DG in-parallel mode.

D. System Simulation Results

A small-scaled tugboat power plant is configured as the target vessel for simulation test. It is a typical hybrid shipboard dc-grid system design, as shown in Fig. 2. It comprises two sets of 200-kW fuel cells, two battery arrays with capacity of 113 kWh and two 410-kW DGs as the back-up power. The detailed system configuration parameters for this dc-based demo vessel are shown in Table I. The simulation tests are performed with two operating modes, respectively: Zero-emission mode and DG in-parallel mode. The test results are shown as follows.

1) *Zero-Emission Mode*: During the zero-emission mode operation, there is no DG online. The battery units are

working under the voltage control scheme to regulate the system dc-grid voltage level. Two battery sets share the load via voltage droop control. Fig. 11 shows the simulation results for the dc-grid hybrid power plant performance under zero-emission operation.

- a) $T = 3-15$ s: The load power increases to 300 kW. The total load demands are supplied by two fuel cell modules and two battery units. Fuel cells follow the power references from the dc/dc converter, while batteries share the remaining required power equally when the voltage droop rate is set as the same value. In this test, voltage droop rate is set as 3%.
- b) $T = 15-30$ s: The propulsion load is dropped. The batteries switch to charging mode to absorb the remaining power from network for future usage.
- c) $T = 30$ s-end: the load is increased to peak, around 400 kW. The batteries discharge and provide the power supply for the peak load demands.

It shows that the dynamic response of fuel cells is relatively slow and it needs the battery systems' support to fulfill the peak ship load. With voltage droop control, the system dc voltage slightly varies according to the different loading conditions. Two fuel cells are operated independently according to their own reference profiles. Usually in real practice, fuel cells can be set to provide the average load demands [36], and the batteries are to enhance the system dynamic performance and boost the power supply during peak load.

2) *DG in-Parallel Mode*: When the system is under DG in parallel mode, the DGs connect to dc network via a

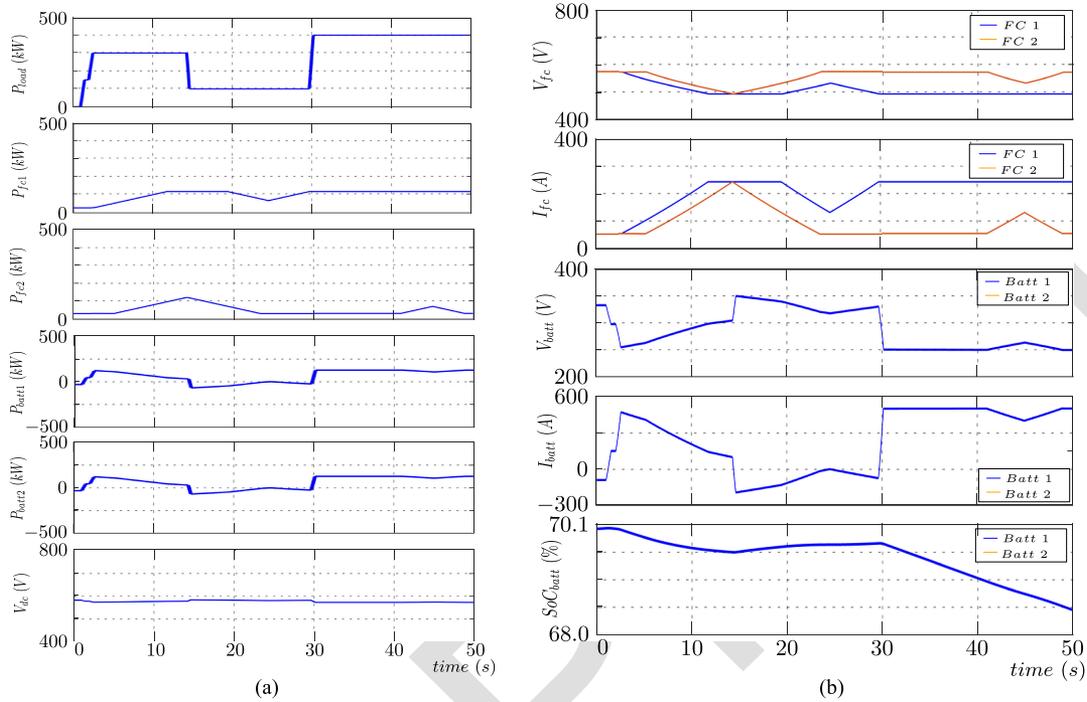


Fig. 11. Simulation results for the dc-based shipboard power grid model for demo vessel. Zero-emission mode. (a) Power split of the hybrid shipboard power plant model. (b) Dynamic behavior of fuel cell devices and batteries.

496 6-pulsed passive rectifier. The online diesel generators
 497 dominate the system dc voltage level naturally by its
 498 terminal ac voltage. In this case, battery systems are switched
 499 to power control scheme which is similar to the fuel
 500 cell control method to provide the output power flow as
 501 reference required to achieve an ideal loading strategy and
 502 optimization control. Fig. 12 shows the simulation results
 503 of the hybrid ship power system under DG in-parallel
 504 mode.

- 505 a) $T = 3 - 15$ s: The load power increases to peak 400 kW.
 506 The fuel cell modules follow the power reference com-
 507 mands from the dc/dc converter. The propulsion load
 508 variations are absorbed and equally shared between
 509 two battery units. The voltage droop rate here is set as
 510 2% for both the DGs so that they are able to share equal
 511 100-kW load throughout the simulation.
 512 b) $T = 15 - 30$ s: The propulsion load is dropped. The
 513 batteries switch to charging mode to absorb the extra
 514 power for future usage.
 515 c) $T = 30$ s - end: The power of the load is shooting up to
 516 the peak again. The batteries discharge and boost the
 517 total power supply.

518 In this simulation test, fuel cells and battery systems provide
 519 the power supply according to the power allocation reference,
 520 and the remaining power demands from the network go to DGs
 521 online. Similar to the zero-emission mode, the batteries function
 522 as the dynamic enhancement to support the system dynamic
 523 performance. To maintain system stability, the generators shall
 524 remain stable outputs, and usually, it can be set as the average
 525 load demands.

526 For both zero-emission mode and DG in-parallel mode sim-
 527 ulation test, the different power sources are integrated into the
 528 common dc-distributed power network and able to satisfy the
 529 load demands. The system power load sharing is under dc
 530 voltage droop combined with power control scheme. During
 531 both of the operation modes, the power reference profile for
 532 the two fuel cells are set differently. This system-level hybrid
 533 power plant model provides the control freedom to interface
 534 with different tertiary level power management algorithms.

535 IV. HIL SETUP

536 Control algorithm testing can be time-consuming, costly, and
 537 potentially unsafe if the test is against a real hardware-based
 538 system. Therefore, software-based plant model is considered
 539 for replacing the real system and to generate the operational
 540 profile of a vessel. A HIL setup is expected to provide real-time
 541 simulation of the shipboard power system behavior and dynam-
 542 ics, as well as the hardwired interfaces with control signals and
 543 communication protocols. HIL test is able to ensure high quality
 544 of the control software. It is a reliable verification and validation
 545 method to prototype the controller before implementing the
 546 control algorithm into a real hardware environment [37].

547 In this study, the shipboard power system plant model is
 548 successfully built and run on a real-time Speedgoat target
 549 machine, which provides real-time simulation of plant model
 550 behavior and dynamics as well as the communication interface
 551 with field-bus protocols. The power and energy optimization
 552 control algorithms are developed and downloaded to a B&R
 553 brand X20CP3586 programmable logic controller (PLC)-based

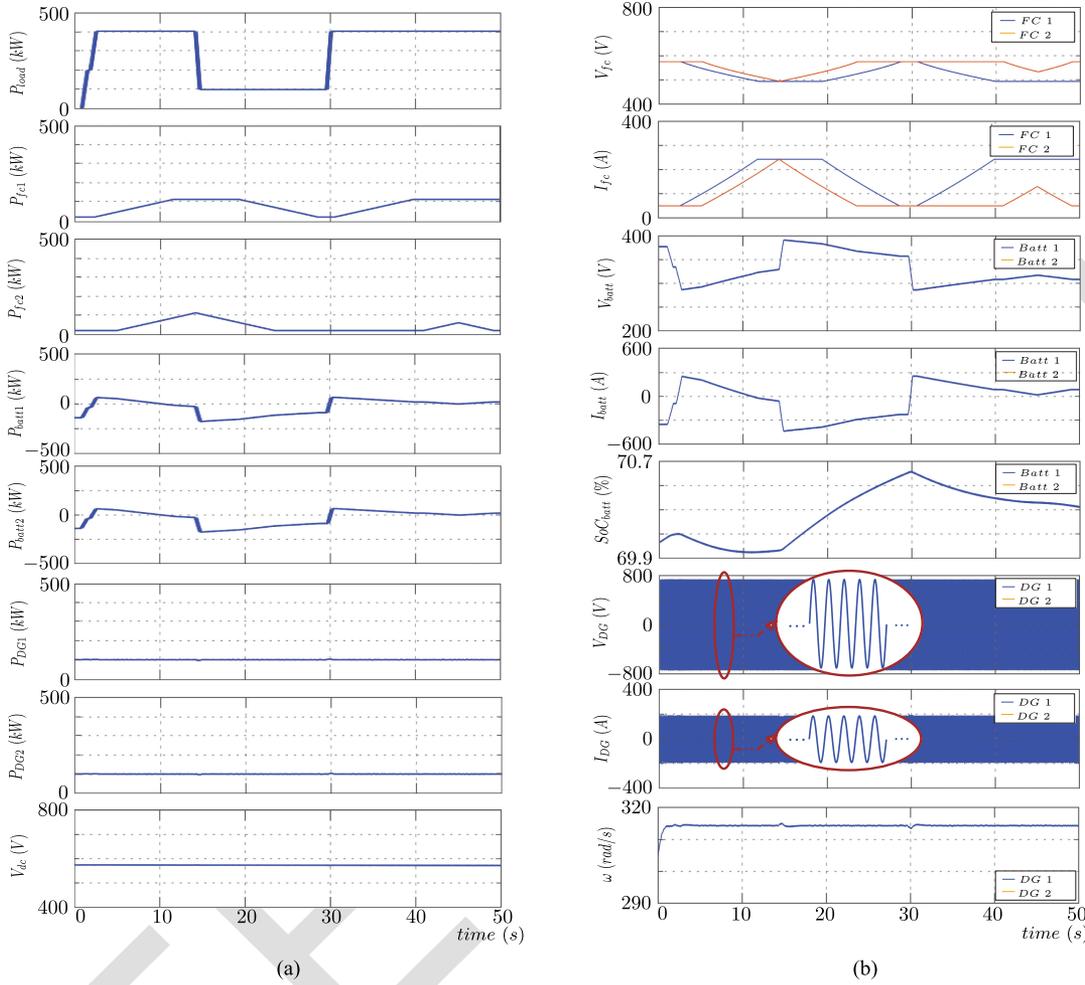


Fig. 12. Simulation results for the dc-based shipboard power grid model for demo vessel. DG in-parallel mode. Frequency of the DG is 50 Hz. (a) Power split of the hybrid shipboard power plant model. (b) Dynamic behavior of fuel cell devices, batteries and DGs.

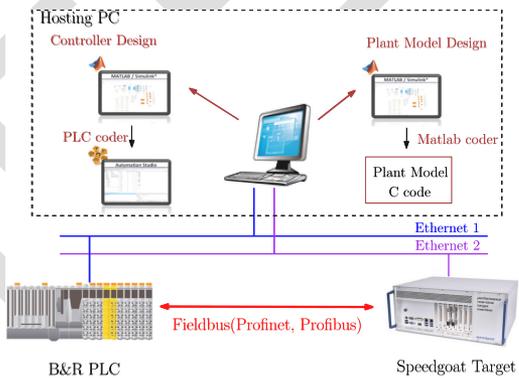


Fig. 13. HIL topology scheme.

554 embedded control system. The available real-time Fieldbus communication protocols between the plant model and controller are
 555 Profinet and Profibus for this setup. In this study, the topology
 556 diagram of the HIL test bed is shown in Fig. 13.
 557

V. SYSTEM MODEL VALIDATION

559 In this study, the mathematical shipboard power system model
 560 and HIL plant are both validated against the full-scale hybrid

power plant system. The actual power grid facilities are located
 561 at ABB collaborated hybrid power laboratory in MARINTEK,
 562 at Norwegian University of Science and Technology, in Trond-
 563 heim, Norway. The main objective of this experimental valida-
 564 tion test is to verify that the response of the modeled shipboard
 565 power system matches with the actual system response. The
 566 DGs and fuel cells with battery energy storage will be validated,
 567 respectively, according to the actual lab setup.
 568

The system configuration of the MARINTEK laboratory is
 569 based on a full-scale hybrid AES with ABB onboard dc-grid
 570 system. The lab arrangement diagram is shown in Fig. 14. The
 571 lab is equipped with two 410-kW diesel engines and variable
 572 speed generator sets, two 30-kW fuel cell devices, plus one
 573 55-kWh battery ESS. Power sources are connected to the dc-bus,
 574 and system loading conditions are simulated with two control-
 575 lable electric motors. Fig. 15 shows the equipment setup in
 576 MARINTEK lab.
 577

In the first validation test, one DG and one 55-kWh battery
 578 system are in use. The validation test results are shown in Fig. 16.
 579 The load profile is designed as four ramp-up cycles. The dynamic
 580 response of a DG is limited by its mechanical characteristic,
 581 and so the battery banks are there to boost the system dynamic
 582 performance to achieve the load demands. In this test, the
 583

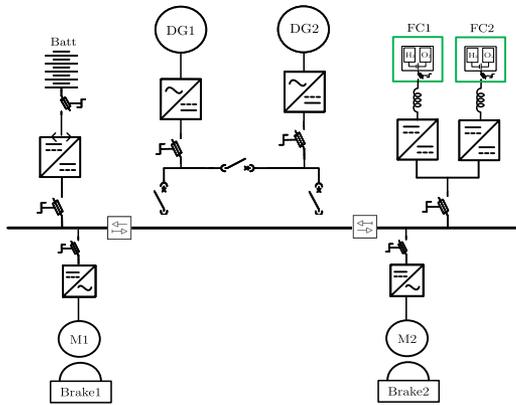


Fig. 14. ABB MARINTEK lab setup diagram, where M1 & M2 are the propulsion motors, and Brake1 & 2 present the controllable thrusters load.



Fig. 15. ABB MARINTEK lab in Trondheim, Norway.

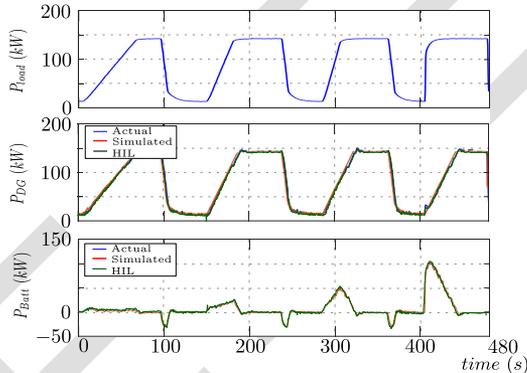


Fig. 16. Validation test I: To compare the system response between simulation model and HIL plant target in DG in-parallel mode.

584 simulated results from the Simulink model can match well with
 585 the actual recorded test. The overall mean average percentage
 586 error between the mathematical plant model and actual system
 587 is around 0.51%, and the average error between HIL plant and
 588 actual system is about 1.82%.

589 In the second validation test, two fuel cell modules and the
 590 same battery system are configured. The test results are shown in
 591 Fig. 17. During the first 100 s, the two fuel cells were powered up
 592 with 9-kW output one after another. The power drawn from the
 593 batteries was reduced with the increasing power supply from
 594 the fuel cells. From the 120 s, the load demands are reduced
 595 below the fuel cell power supply so that the batteries are charged
 596 by the fuel cells. Between 100 to 480 s, the fuel cells provide
 597 the constant power output, while the batteries absorb the load

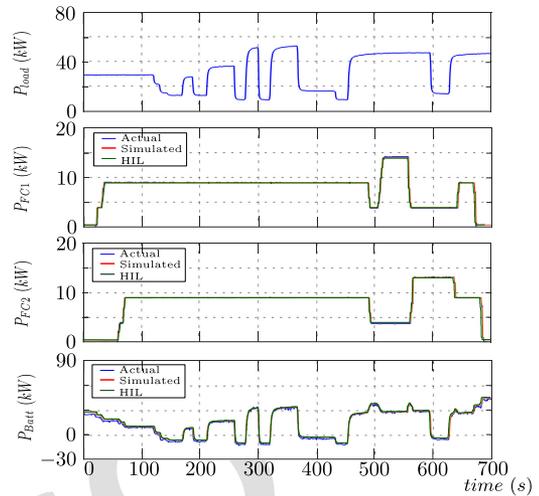


Fig. 17. Validation test II: To compare the system response between simulation model and HIL plant target in Zero-emission mode.

598 variations. From the 480 s onward, the power references for fuel
 599 cells are altered so that the batteries provide the remaining power
 600 to the load. The mean average percentage error is about 1.98%
 601 between the mathematical model and the actual system, while it
 602 is about 1.47% between HIL plant and the actual system.

603 To compare the test results, it can be seen that the system
 604 simulation and HIL results compare well with the experimental
 605 values recorded from the site. In addition, the transient rise time
 606 and steady-state power output for all DG, fuel cells, and battery
 607 power sources for both mathematical model and HIL plant can be
 608 in line with the actual system responses. Therefore, the proposed
 609 hybrid shipboard power plant model can be used to represent the
 610 physical system behavior.

VI. CONCLUSION

611
 612 In this article, ad-c-distributed hybrid shipboard power system
 613 configuration has been studied. A system-level fuel cell-fed
 614 hybrid shipboard power grid system with dc distribution has
 615 been modeled. In addition, an HIL plant model platform has
 616 been set up and tested in a lab environment. Finally, both
 617 the system mathematical model and the HIL plant have been
 618 validated against a full-scale hybrid shipboard power system.
 619 This research work provides an essential platform and solid
 620 foundation to apply optimization control of power and energy
 621 management strategy in future work.

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622
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DC-Distributed Power System Modeling and Hardware-in-the-Loop (HIL) Evaluation of Fuel Cell-Powered Marine Vessel

Wenjie Chen¹, Student Member, IEEE, Kang Tai, Michael Lau, Ahmed Abdelhakim², Senior Member, IEEE, Ricky R. Chan, Alf Kåre Ådnanes, and Tegoeh Tjahjowidodo

Abstract—Environmentally driven regulations are significantly affecting shipping in recent years, where the shipbuilding industry is required to comply with upcoming restrictions concerning polluting emissions. The all-electric ship (AES) is one of the most promising technologies for complying with the increasingly strict environmental regulations, improving fuel efficiency, and enhancing system dynamic performance. In this study, the dc-distributed power grid of an AES integrated with fuel cells and batteries has been configured using extensive electrification technology, where the system-level shipboard power plant has been modeled with the average modeling method. The model not only incorporates the hybrid power source integration but also the primary and secondary power management as a whole. In addition, a hardware-in-the-loop (HIL) has been set up to replicate the real-time system behavior, which is essential for the verification of any optimal power management control algorithms to be developed in future work. Finally, both the mathematical and real-time models are validated against the full-scale hybrid shipboard power system.

Index Terms—All-electric ship, fuel cell system (AES), hardware-in-the-loop (HIL), hybrid shipboard power microgrid, renewable energy sources.

I. INTRODUCTION

MARITIME transport has played a dominant role in the global trade system for centuries, and it is expected that

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international maritime trade will expand at continuous annual growth for years to come [1]. Recently, environmental sustainability has become a significant policy concern in global maritime transport [2]. With this concern, the shipbuilding industries are compelled to abide by rules on polluting emissions imposed by International Maritime Organization (IMO) and national legislation [3]. To achieve high energy efficiency and low emission, it is believed that the all-electric ship (AES) will be one of the most promising technologies to comply with the environmental regulations [4]. In particular, vessel types such as tug boats, river pushers, and ferries will be the early adopters as they usually operate near coasts or in inner rivers, which belong to emission control areas (ECAs). Other vessel types such as container vessels may need more time before the AES design can be adopted [5].

With the development of energy storage technologies, new energy sources can be integrated into the shipboard power network. Batteries, as the most mature energy storage system (ESS) device, have been widely utilized in different applications in marine vessels. However, due to energy density limitation, battery systems only function as a main power supply for small-scale vessels in short distance shipping segment or as an auxiliary power source onboard of large vessels to enhance the dynamic performance or short-term zero-emission operation. In order for clean energy sources to be a feasible solution for larger ocean-going ship types, such as cruise ships and RoRo/RoPax (the vessels built for freight vehicle transport with passenger accommodation), fuel cell systems have been introduced. It is expected that fuel cell systems would replace diesel engines as the main power source in the future.

The hybrid shipboard power system integrated with different types of power sources brings in a lot of benefits, but it also increases the complexity of the system structure and power management control. To get deep understanding of the hybrid power system and improve the current control algorithms, a comprehensive system-level hybrid power plant model is needed as a foundation. The expected system model should be able to perform the actual shipboard power plant responses under different operating modes and loading profiles.

A few power system configurations and modeling methodologies have been published in the recent literature. In [6] and [7], land-based hybrid dc microgrid configurations have

72 been introduced. The authors in [8] have proposed a diesel-
 73 generator-based marine power network equipped with a battery
 74 ESS, while a fuel cell-based marine system model has been
 75 developed in [9], where the authors are mainly focusing on
 76 the fuel cell chemical reaction and fuel processing. The above
 77 references focus only on certain parts of the power grid rather
 78 than the observation of the dynamics and performance of the
 79 complete system.

80 It is worth noting that shipboard power systems have their
 81 own special configurations and requirements [10]. The main
 82 difference with land-based power system is that shipboard power
 83 system is an isolated microgrid with short distances from gen-
 84 erated power to the electric propulsion load [11]. The average
 85 short-circuit levels and forces are quite high so that it must be
 86 dealt with some special manners to protect the power equipment.
 87 During cruising, shipboard power system is impacted a lot by the
 88 environment, such as marine weather and unexpected thrusters
 89 load variations, therefore the system stability and reliability is
 90 crucial for marine vessels. Extra efforts in system reconfigurabil-
 91 ity and redundancy are required to prevent single point failure
 92 according to the marine classification rules. Compared to EV
 93 power system, the power scale of the shipboard power network
 94 is much bigger and complicated. Due to its particularity of the
 95 shipboard power system [12], there is the urge to build its own
 96 power plant model rather than reusing other existing models.

97 In addition, system-level power plant model is not only to inte-
 98 grate different power sources but also to implement primary and
 99 secondary level power management to coordinate power load
 100 sharing, frequency, and voltage control. A previous work by the
 101 authors [13] modeled an ac-distributed shipboard hybrid power
 102 system. However, in the past decade, dc-based distribution sys-
 103 tems have gained popularity due to their advantages in many
 104 aspects [14]. A system-level dc-based shipboard power model
 105 is necessary to simulate the vessel operating performances. The
 106 comprehensive and reliable power plant model is the foundation
 107 and precondition for system studies and further research.

108 Compared to the existing studies on hybrid power network
 109 modeling problems, the main contributions of this article are
 110 highlighted as follows.

- 111 1) A dc-distributed system-level shipboard power system
 112 model is built, integrating the mathematical model of differ-
 113 ent power sources, as well as the primary and secondary
 114 power management control. All the power sources are res-
 115 izable to configure the new shipboard systems according
 116 to the varied ship design specifications.
- 117 2) A hardware-in-the-loop (HIL) platform is set up in lab
 118 environment as well. HIL helps to perform the system
 119 real-time states. It is a useful tool for prototyping the power
 120 and energy management system (PEMS) controller and
 121 communication interface validation without involving any
 122 hardware devices setup.
- 123 3) Both the mathematical and real-time HIL model have
 124 been validated against the real hardware full-scale hybrid
 125 shipboard power system.

126 The rest of this article is organized as follows: The hybrid
 127 shipboard power system configuration is discussed in Section II.
 128 A dc-based power distribution model and the simulation results

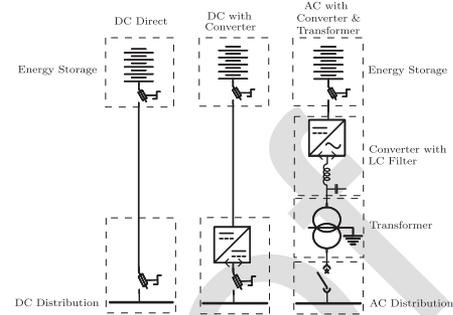


Fig. 1. ESS integration scheme for both dc and ac distribution system.

are presented in Section III. Section IV evaluates the system
 plant model HIL setup. And the validation tests are performed
 in Section V. Finally, Section VI concludes this article.

II. SHIPBOARD POWER SYSTEM CONFIGURATION

In this section, the functions of several integration schemes
 will be introduced in detail. The advantages and the system
 configuration of dc-distributed shipboard power system will also
 be discussed.

A. Advantages of DC-Based Shipboard Power System

Currently, the majority of commercial marine vessels have a
 diesel or gas engine powered plant with ac distribution, while
 the dc-distributed shipboard system has drawn much attention
 over the past decade. In both ac and dc-distributed systems, ESS
 integration is a trend for different functionalities [15].

Energy storage, such as batteries will help the slow power
 devices level out load variations from the thrusters and other
 heavy consumers. Besides, the clean energy sources are able to
 provide the main power supply of the ship operation in order
 to achieve zero-emission operation. There are several methods
 to integrate the ESS into the shipboard network [16], which is
 shown in Fig. 1.

According to this integration scheme, it can be seen that the
 ESS is much easier to connect to a dc-distributed network than an
 ac system. The storage devices can be directly connected to the
 dc network. This direct-on-line (DOL) ESS gives an efficiency
 increase compared to the ESS with dc/dc converter, but at the
 cost of less control features to be achieved. On the other hand,
 ESS is also possible to be extended to an ac-distributed system
 through a dc/ac converter with LCL filter and transformer [17].
 The LCL filter is used herein to improve the power quality and
 reduce the total harmonic distortion (THD) level of the main
 power network. While the shielded transformer is utilized to
 solve the electromagnetic interference (EMI) issue, it also blocks
 the circulating current within the power electronic converters
 such as IGBT or MOSFET. In addition, the dc/ac converters also
 need to monitor the amplitude, frequency, and phase shift of the
 power network voltage on ac side to achieve the synchronization
 operation [13]. It is clear that the ac-based power configuration
 is quite complex not only in system configuration but also in
 power management control [14]. Furthermore, it also increases

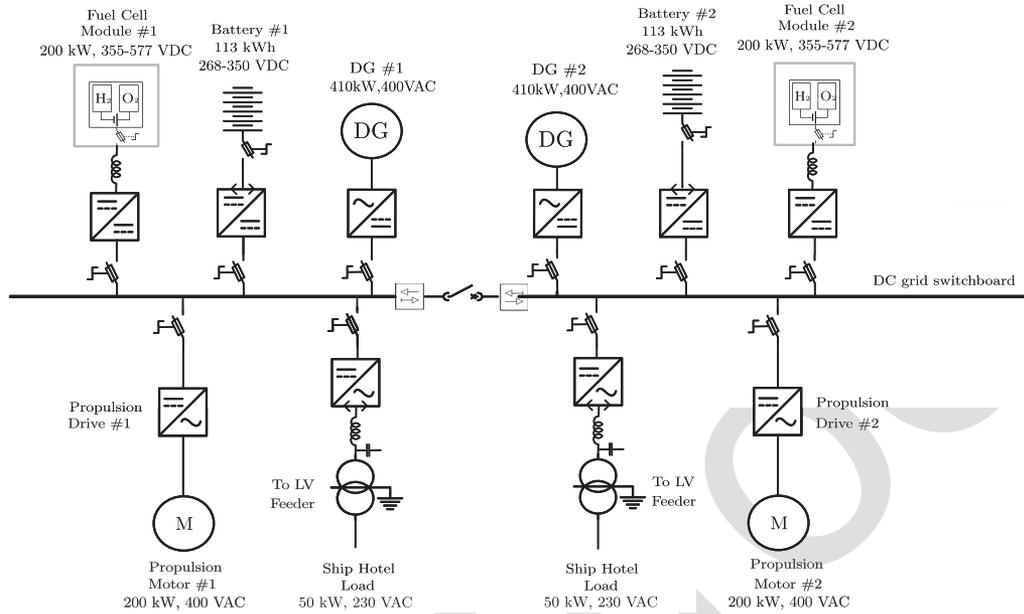


Fig. 2. Typical dc-based shipboard power system single line diagram.

the dimensions of the equipment, where the space onboard a ship is limited.

Safety is another important consideration for implementing the dc-based distribution. A dc-configured system is inherently simpler than an ac system, which means that it is easier to predict fault scenarios and devise adequate protection against them. The generators can be connected to the dc distribution system in a short time because it is not necessary for synchronization. It has considerable potential to improve the stability, efficiency, and performance of future dc-based ship power systems [18].

In addition to the benefits described above, there are numerous other benefits with dc distribution. Some of these are summarized as follows.

- 1) *Voltage Distortion*: Harmonic distortion is inherent in ac systems with frequency converters, while being less of a concern with dc distribution.
- 2) *Electrical Efficiency*: In the process of going from ac-to-dc distribution, the system efficiency is improved by 0.5–1% [19].
- 3) *Ease of Control*: It is not required to control the frequency and reactive power of the network. The control system is simplified and only needs to overlook the dc voltage.

Overall, the integration of a hybrid ac-distributed shipboard power is complex and costly [14]. Multiple layers of the power conversion and transmission increase the system losses and probability of equipment failure. It is more suitable for some retrofit projects, which are planned for extending a single ESS from an existing ac distribution network. Due to the advantages of dc-distribution power system, this study focuses on dc-based shipboard power plant modeling.

B. Hybrid Shipboard Power System Configuration

For a hybrid dc-distributed shipboard power system, the energy storage devices can be connected to the system. A generator set driven by a combustion diesel engine is possible to be integrated into the dc power network. Usually, synchronous generators are used and rectified (ac to dc converter) before the power is transmitted to the dc bus. Propulsion and other ship service loads are regulated by dc/ac inverters. The typical configuration of an all-electric dc hybrid ship power and propulsion system is shown in Fig. 2. In this study, two fuel cell modules are function as the main power supply to achieve zero emission operation. Two battery banks handle the ship load variations. And two diesel gen sets (DG) are the backup power in case the fuel cell devices or the hydrogen fuel are not available under emergency condition.

The main switchboards are usually split into two sections or more to obtain the redundancy requirements of the vessel according to the marine class rules and regulations. This kind of structure is not only an intrinsic advantage of dc distribution but it also has the ability to supply power to the load continuously even under certain fault conditions [16], such as short-circuit. During normal operation, the dc-grid switchboard are usually connected (close bus tie), which gives the best flexibility in the configuration of the power generation. The optimal number of the power devices can be connected to the power network according to the load transients to achieve certain control criteria.

III. FUEL CELL-FED DC DISTRIBUTION SYSTEM MODELING

In this section, a system-level dc-based hybrid shipboard power plant will be modeled. The power network includes the fuel cells, batteries, and DGs as the energy sources. The essential components of the dc-grid power system are the power

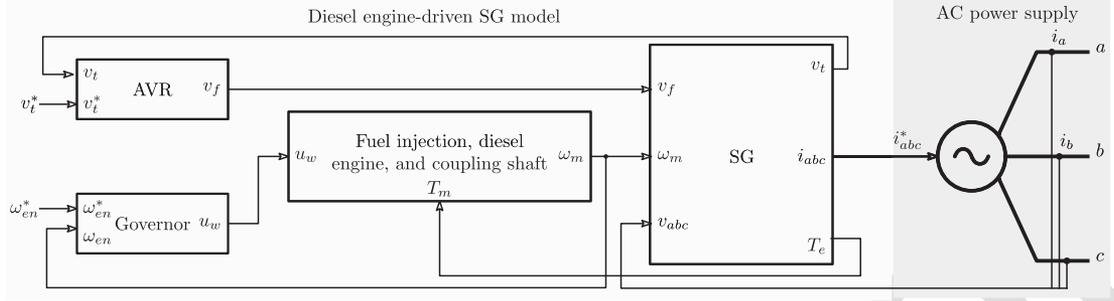


Fig. 3. General block diagram of a diesel engine-driven synchronous generator.

electronics devices. An ESS dc/dc converter and a 6-pulse ac/dc rectifier for DG integration will be modeled. The propulsion load and ship hotel loads of the vessel will be represented with controllable current sources. Power management, including load power sharing, dc voltage coordination, and power regulation will be achieved as well.

A. Energy Sources Modeling

1) *Diesel Gen-Set Model*: Fig. 3 shows the general configuration of a diesel engine-driven synchronous generator system. The system consists of four main sections: Fuel injection and diesel engine, governor, synchronous generator (SG), and automatic voltage regulator (AVR) [20]. The mathematical models of each part of diesel generator are introduced as follows.

a) *Fuel injection and diesel engine* can be presented by a time delay and the coupling shaft model, as in [21] and [22]. The simplified model is given by

$$\frac{T_m(s)}{u_\omega(s)} = \frac{K_{en} \cdot e^{-t_d \cdot s}}{t_e s + 1} \quad (1)$$

$$J_{eq} \cdot \frac{d\omega_m}{dt} = T_m - T_e - k_{f_{eq}} \cdot \omega_m \quad (2)$$

where ω_m is the mechanical speed, u_ω is the control signal from the speed governor, T_m is the mechanical torque developed by the engine, J_{eq} is the equivalent inertia of the entire system, $k_{f_{eq}}$ is the equivalent friction coefficient, K_{en} is the engine gain, t_d is a delay representing the time elapsed from the fuel injection until the torque is developed at the engine shaft, and t_e is the time constant of the fuel injection.

b) *Governor* is responsible for regulating the engine speed to maintain it within the allowable range during the different load conditions, resulting in constant output frequency or within the desired limits. This governor also has the structure of a PI controller with the droop function implemented. This model can be given by

$$u_\omega = (\omega_{en}^* - \omega_{en} - k_{dr, \text{freq}} \cdot u_\omega) \left(K_{P_\omega} + \frac{K_{I_\omega}}{s} \right) \quad (3)$$

where K_{P_ω} and K_{I_ω} are the proportional and integral gains of the governor PI controller, and ω_{en} is the engine nominal speed and $k_{dr, \text{freq}}$ is the speed droop

gain that equals $(m_{dr} \cdot \omega_{en})$, and m_{dr} is the static droop slope.

c) *Synchronous generator* is considered without damper windings for the sake of simplifying the model. And its d - q rotating reference frame is given by

$$V_d = -r_s \cdot i_d + L_q \cdot \omega_e \cdot i_q + L_d \cdot \frac{di_d}{dt} + M_{sf} \cdot \frac{di_f}{dt} \quad (4)$$

$$V_q = -r_s \cdot i_q - L_d \cdot \omega_e \cdot i_d + L_q \cdot \frac{di_q}{dt} + M_{sf} \cdot \omega_e \cdot i_f \quad (5)$$

$$V_f = r_f \cdot i_f + L_f \cdot \frac{di_f}{dt} - M_{sf} \cdot \frac{di_d}{dt} \quad (6)$$

$$T_e = (L_d - L_q) \cdot i_d \cdot i_q + M_{sf} \cdot i_q \cdot i_f \quad (7)$$

where V_d , V_q , i_d , and i_q are the output voltages and currents in the d - q rotating reference frame, respectively, V_f and i_f are the field excitation voltage and current, respectively, r_s and r_f are the stator winding and field winding internal resistances, respectively, L_d and L_q are the stator inductance in the d -axis and q -axis, respectively, L_f is the inductance of the field, M_{sf} is the mutual inductance between the field winding and the d -axis stator winding, and T_e is the electromagnetic torque. It is worth noting that, the following assumptions have been considered: The stator windings are symmetrical, a uniform sinusoidal distribution along the air gap, the permanence of the magnetic paths on the rotor is independent of the rotor positions, and the saturation and the hysteresis effects are neglected.

d) *Automatic voltage regulator* has the responsibility of controlling the terminal voltage of the synchronous generator under different load conditions. This AVR is modeled as a first-order system, representing a power converter controlled by a PI controller [20], that is given by

$$V_f = \frac{k_{conv}}{t_{conv} s + 1} \cdot (U_v - k_{dr, v} V_f) \quad (8)$$

$$U_v = \left(K_{P_v} + \frac{K_{I_v}}{s} \right) \cdot (V_t^* - V_t) \quad (9)$$

where K_{P_v} and K_{I_v} are the proportional and integral gains of the AVR PI controller, k_{conv} and t_{conv} are the

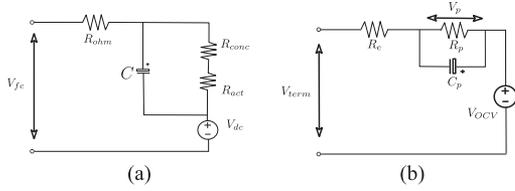


Fig. 4. ESS power source modeling. (a) Proton-exchange membrane (PEM) fuel cell and its equivalent circuit. (b) Lithium-ion battery and its equivalent circuit.

converter gain and time constant, and V_t^* and V_t are the reference and measured rms output line voltage. $k_{dr,v}$ is the dc voltage droop rate for dc-distributed power system.

- 2) *Fuel Cell Model*: There are a few different types of fuel cell technologies. The proton-exchange membrane (PEM) fuel cell is the most mature technology over other types of fuel cell. With its relatively low cost and high energy efficiency features, PEMFC is the most widely used in marine applications [23]. Fig. 4(a) shows a single PEM fuel cell model and its equivalent circuit [24]. The fuel cell losses are mainly divided into three categories: Activation loss R_{act} , concentration loss R_{conc} , and ohmic loss R_{ohm} [25]. The voltage of a PEMFC elementary cell can be written as follows [26]:

$$V_{cell} = E_{nemst} - V_{act} - V_{conc} - V_{ohm} \quad (10)$$

$$V_{fc} = N_{fc} \times V_{cell} \quad (11)$$

where V_{cell} is the voltage of a PEMFC elementary cell, E_{nemst} is the equilibrium voltage, V_{act} is the activation overpotential, V_{conc} is the concentration overpotential, V_{ohm} is the ohmic overpotential, V_{fc} is the voltage of PEMFC stack, and N_{fc} is the number of cells in series.

In this study, the PEMFC is configured with Matlab/Simulink Simscape library.

- 3) *Battery Model*: To balance the cost, energy density, power density, safety performance, and life span, Li-ion battery is the most applied battery type in marine ESS applications [27]. Simple models based on the equivalent circuit of a resistor in series with a parallel RC circuit (polarization circuit) [28] have been used to determine the dynamic characteristics of the battery, as shown in Fig. 4(b).

Open-circuit voltage (OCV) is defined differently for discharging and charging at given state-of-charge (SoC) due to the hysteresis effect [29]. The total voltage drop from OCV is then modeled here based on the first-order equivalent circuit to account for the internal resistance and polarization (OCV relaxation).

$$V_{term} = V_{ocv} - IR_e - V_p \quad (12)$$

where V_{term} is the terminal voltage of the cell, V_{ocv} is the cell OCV, I is the current rate of the cell (positive for discharging and negative for charging), R_e is the electrical resistance (responsible for instantaneous voltage drop together with C_p , after the application of load) and V_p is the voltage drop across the polarization circuit.

At any instant, V_p across the parallel RC circuit is represented by the voltage drop across R_p or C_p . So, the equation can be arranged by

$$\frac{V_p(t)}{R_p} + C_p \frac{dV_p(t)}{dt} = I(t) \quad (13)$$

where I_{Rp} is the current across polarization resistance, and I_{Cp} is the current across polarization capacitance. The sum of I_{Rp} and I_{Cp} gives the total circuit current I . Then perform Laplace transform by

$$V_p(s) = I(s) \frac{R_p}{1 + sR_pC_p} \quad (14)$$

so that the steady-state response of terminal voltage $V_{term s-s}$ to a given current by

$$V_{term s-s} = V_{ocv} - I(R_e + R_p) = V_{ocv} - IR_{ti} \quad (15)$$

where $R_{ti} = R_e + R_p$ is the total internal resistance of a cell.

SoC is another critical parameter for the battery system. SoC is intrinsically divided into two types: Instantaneous SoC (SoC_{inst}) and nominal SoC (SoC_{nom}), both based on Ah counting. Nominal SoC is derived from nominal capacity based on standard discharging conditions as suggested by the manufacturer, while instantaneous SoC is based on the available instantaneous discharge capacity given by

$$SoC_{nom}(t_i) = SoC_{nom}(t_{i-1}) - \frac{I_{step}(t_{i-1})t_{step}}{C_{nom}} \quad (16)$$

$$SoC_{inst}(t_i) = \frac{(1 - SoC_{nom}(t_{i-1}))C_{nom}}{C_{step\ avail}(t_{i-1})} - \frac{I_{step}(t_{i-1})t_{step}}{C_{step\ avail}(t_{i-1})} \quad (17)$$

where, I_{step} is the current at a given time step, t_{step} is the span of time under consideration (sampling time of BMS/controller), and $C_{step\ avail}$ is the capacity available for I_{step} . Instantaneous SoC is useful in indicating the capacity available at a given time step and operating conditions, especially higher discharge rates.

In this study, the Matlab/Simulink Simscape battery stack model is utilized.

B. Power Electronic Converter Modeling

A dc/dc converter (for fuel cell and battery control), and an ac/dc rectifier (for diesel generator control) are modeled in this section.

- 1) *DC/DC Converter Model*: Both of the fuel cells and the battery systems are powered by dc/dc converter. The dc/dc converter regulates the output voltage and power flow from the storage device [8]. Different from the battery bidirectional control [31], fuel cells only require a single direction power flow to provide the power to dc link. It can be represented as the following equation:

$$\frac{V_{dc}}{V_{ESS}} = \frac{1}{1 - D} \quad (18)$$

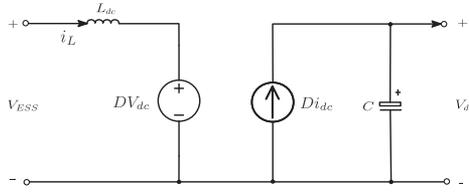


Fig. 5. DC/DC converter average model.

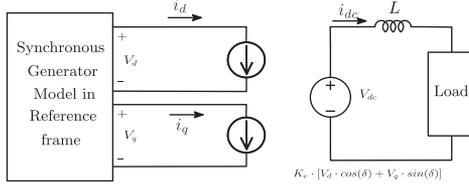


Fig. 6. Average model of 6-pulse ac/dc rectifier, where V_d , V_q , I_d , I_q are generator's output voltage and current in $dq - axes$, δ is the angle between V_d and V_q , coefficient K_v is the ratio between the magnitude of rectifier output voltage V_{dc} and its input voltage V_d , V_q , and δ .

where D is the duty cycle of dc/dc converter.

There are two methods to model the power electronic converters, i.e., electromagnetic transient (EMT) model and average model. EMT model is straightforward and can represent the modulation process of the power electronic device. However, the simulation speed will be slow down due to the usage of high-frequency carriers during pulsewidth modulation (PWM). EMT simulation method is more suitable for system stability analysis studies, but not suitable to evaluate system performance throughout the operational profile which can last in the order minutes or even hours. For that reason, the average model is selected to configure the shipboard power plant model here. The dc/dc converter average model is shown in Fig. 5.

2) *AC/DC Rectifier Model*: The DGs are connected into dc bus via a 6-pulse diode rectifier. The average model of a diode bridge loaded synchronous generator was used here to simplify the calculation of the average values of generator $d-q$ variables to the dc voltage and current at the bridge output [32], as shown in Fig. 6.

The voltage and current at the dc terminal can be defined as proportional to the amplitudes of the first harmonics of generator phase voltage and current, respectively [33]

$$V_{dc} = K_v \sqrt{V_d^2 + V_q^2} \quad (19)$$

$$i_{dc} = K_i \sqrt{i_d^2 + i_q^2} \quad (20)$$

where V_{dc} is dc voltage output from the rectifier. V_d , V_q , i_d , and i_q are the voltage and current of generator in dq -axes. K_v is the generator voltage constant, and K_i is the generator current constant.

C. System Control Philosophy

For existing hybrid shipboard power management is usually arranged in line with the hierarchical control framework. The basic structure is shown in Fig. 7.

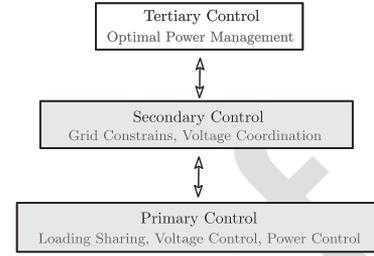


Fig. 7. Hierarchical control framework for shipboard power management, where the primary and secondary control have been applied.

In hierarchical control, the primary control level is to handle the load sharing among the hybrid power sources. The secondary control level includes the grid voltage coordination and to ensure bus signals at their operating ranges. The tertiary control level is used to achieve optimal operation with intentional objectives [34]. In this study, the primary and secondary level power management control have been achieved, while the tertiary control will be considered in future's research work.

The basic load sharing mechanism in dc-distributed power network is dc voltage droop control. Basically, with the voltage droop control, all the power sources share the load based on the dc bus voltage in the similar way as with frequency droop control in ac-distributed system. The load sharing function is realized by adjusting the voltage droop rate. In Fig. 8, it shows the load sharing results of a dc power network with three different droop rate setting. A steeper droop angle will cause one consumer to take less load than one with a more flat droop angle given the same voltage set point.

DG system is integrated into the dc network via 6-pulse diode rectifier, and the output voltage is controlled by AVR. The dc voltage set point is defined as the nominal dc voltage of the rectifier. Droop increases with increasing generator speed. Usually, the dc voltage droop rate is not larger than 5% [35].

For the energy storage power sources, such as fuel cells and batteries, the dc voltage droop control is implemented in the power electronic dc/dc converters. The block diagram of the voltage control loop is as shown in Fig. 9. It is a dual control loop, with the current control as the inner loop with the voltage control as the outer loop. $K_{dr,v}$ is the voltage droop rate for load sharing with other power sources.

The dc voltage droop control is simple and reliable. But due to the complexity of hybrid dc-grid system configuration, it is difficult to fine-tune the whole power system with one single parameter to maintain the performance of each power source, and ensure the system stability at the same time. It is also not possible to achieve strategy loading or even optimized operation. Therefore, the power control of the ESS devices is expected. The dc/dc converter power control loop with PI controllers is shown in Fig. 10. It is a single current control loop, where the current reference is given by the quotient of reference power and dc voltage.

In this study, dc voltage droop combined with power control is implemented as primary power management scheme. The system voltage coordination rules and constrains have been set in power electronic converter models as the secondary level of

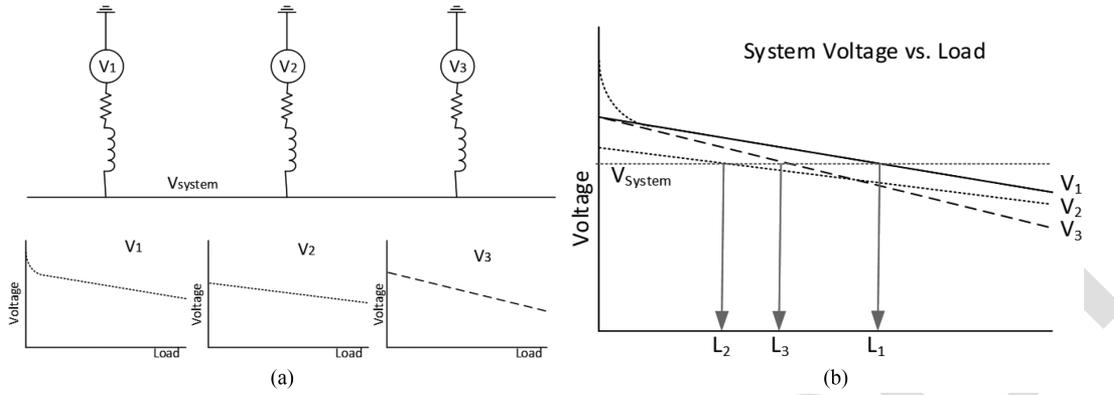


Fig. 8. Load sharing with 3 different droop curves. (a) Integrated dc-distributed system with three power sources and their droop rates. (b) System voltage and load sharing scheme.

TABLE I
PARAMETERS FOR THE DC-BASED DEMO VESSEL

Diesel-generator		Fuel cell [30]		Batteries	
Rated power (P_{gen})	410 kW	Rated power (P_{fc})	2*100 kW	Capacity (E_{batt})	113 kWh
Nominal voltage (V_{abc})	400 VAC	Min. power (P_{fc_min})	20 kW	Max. charge voltage (V_{batt_max})	350VDC
Nominal current (I_{abc})	592 A	Operating voltage (V_{fc})	355-577 VDC	Nominal voltage (V_{batt})	300 VDC
Frequency (F_n)	50 Hz	Rated current (I_{fc})	2*257 A	Min. voltage (V_{batt_min})	268 VDC
Speed (Spd)	1500 rpm	Max. Temperature (T_{max})	70°C	Max. discharge current (I_{batt_disch})	250 A
Inertia (J_{ge})	10 kgm ²	Fuel Type	Hydrogen	Max. charge current (I_{batt_ch})	200 A
Power factor (pf)	0.9	H ₂ pressure	3.5-5 Barg	Operating state of charge range (SoC)	20%-80%

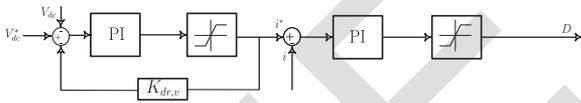


Fig. 9. DC/DC converter control scheme I - Voltage control loop.

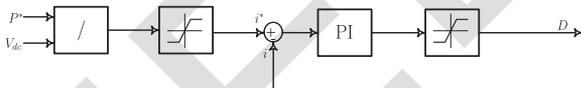


Fig. 10. DC/DC converter control scheme II - Power control loop.

power management. The ESS device output power or voltage shall be regulated according to the reference command if the corresponding control scheme is selected. Two different operation modes will be studied in this article: Zero-emission mode and DG in-parallel mode.

D. System Simulation Results

A small-scaled tugboat power plant is configured as the target vessel for simulation test. It is a typical hybrid shipboard dc-grid system design, as shown in Fig. 2. It comprises two sets of 200-kW fuel cells, two battery arrays with capacity of 113 kWh and two 410-kW DGs as the back-up power. The detailed system configuration parameters for this dc-based demo vessel are shown in Table I. The simulation tests are performed with two operating modes, respectively: Zero-emission mode and DG in-parallel mode. The test results are shown as follows.

1) *Zero-Emission Mode*: During the zero-emission mode operation, there is no DG online. The battery units are

working under the voltage control scheme to regulate the system dc-grid voltage level. Two battery sets share the load via voltage droop control. Fig. 11 shows the simulation results for the dc-grid hybrid power plant performance under zero-emission operation.

- $T = 3-15$ s: The load power increases to 300 kW. The total load demands are supplied by two fuel cell modules and two battery units. Fuel cells follow the power references from the dc/dc converter, while batteries share the remaining required power equally when the voltage droop rate is set as the same value. In this test, voltage droop rate is set as 3%.
- $T = 15-30$ s: The propulsion load is dropped. The batteries switch to charging mode to absorb the remaining power from network for future usage.
- $T = 30$ s-end: the load is increased to peak, around 400 kW. The batteries discharge and provide the power supply for the peak load demands.

It shows that the dynamic response of fuel cells is relatively slow and it needs the battery systems' support to fulfill the peak ship load. With voltage droop control, the system dc voltage slightly varies according to the different loading conditions. Two fuel cells are operated independently according to their own reference profiles. Usually in real practice, fuel cells can be set to provide the average load demands [36], and the batteries are to enhance the system dynamic performance and boost the power supply during peak load.

2) *DG in-Parallel Mode*: When the system is under DG in parallel mode, the DGs connect to dc network via a

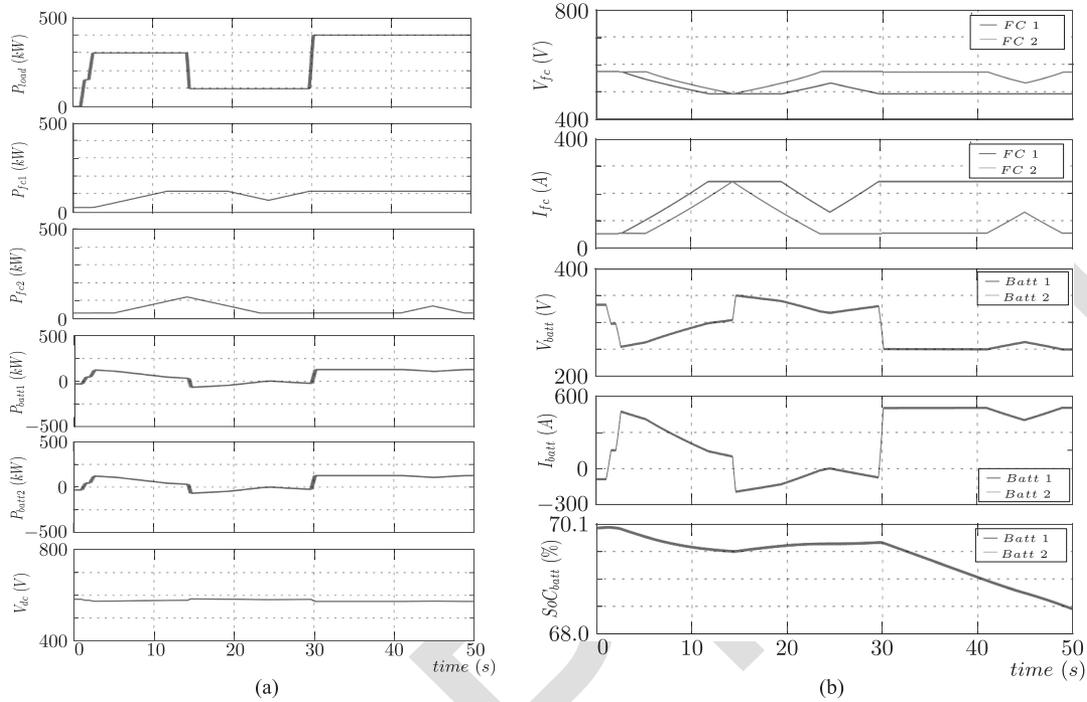


Fig. 11. Simulation results for the dc-based shipboard power grid model for demo vessel. Zero-emission mode. (a) Power split of the hybrid shipboard power plant model. (b) Dynamic behavior of fuel cell devices and batteries.

496 6-pulsed passive rectifier. The online diesel generators
 497 dominate the system dc voltage level naturally by its
 498 terminal ac voltage. In this case, battery systems are switched
 499 to power control scheme which is similar to the fuel
 500 cell control method to provide the output power flow as
 501 reference required to achieve an ideal loading strategy and
 502 optimization control. Fig. 12 shows the simulation results
 503 of the hybrid ship power system under DG in-parallel
 504 mode.

- 505 a) $T = 3 - 15$ s: The load power increases to peak 400 kW.
 506 The fuel cell modules follow the power reference com-
 507 mands from the dc/dc converter. The propulsion load
 508 variations are absorbed and equally shared between
 509 two battery units. The voltage droop rate here is set as
 510 2% for both the DGs so that they are able to share equal
 511 100-kW load throughout the simulation.
 512 b) $T = 15 - 30$ s: The propulsion load is dropped. The
 513 batteries switch to charging mode to absorb the extra
 514 power for future usage.
 515 c) $T = 30$ s - end: The power of the load is shooting up to
 516 the peak again. The batteries discharge and boost the
 517 total power supply.

518 In this simulation test, fuel cells and battery systems provide
 519 the power supply according to the power allocation reference,
 520 and the remaining power demands from the network go to DGs
 521 online. Similar to the zero-emission mode, the batteries function
 522 as the dynamic enhancement to support the system dynamic
 523 performance. To maintain system stability, the generators shall
 524 remain stable outputs, and usually, it can be set as the average
 525 load demands.

526 For both zero-emission mode and DG in-parallel mode sim-
 527 ulation test, the different power sources are integrated into the
 528 common dc-distributed power network and able to satisfy the
 529 load demands. The system power load sharing is under dc
 530 voltage droop combined with power control scheme. During
 531 both of the operation modes, the power reference profile for
 532 the two fuel cells are set differently. This system-level hybrid
 533 power plant model provides the control freedom to interface
 534 with different tertiary level power management algorithms.

535 IV. HIL SETUP

536 Control algorithm testing can be time-consuming, costly, and
 537 potentially unsafe if the test is against a real hardware-based
 538 system. Therefore, software-based plant model is considered
 539 for replacing the real system and to generate the operational
 540 profile of a vessel. A HIL setup is expected to provide real-time
 541 simulation of the shipboard power system behavior and dynam-
 542 ics, as well as the hardwired interfaces with control signals and
 543 communication protocols. HIL test is able to ensure high quality
 544 of the control software. It is a reliable verification and validation
 545 method to prototype the controller before implementing the
 546 control algorithm into a real hardware environment [37].

547 In this study, the shipboard power system plant model is
 548 successfully built and run on a real-time Speedgoat target
 549 machine, which provides real-time simulation of plant model
 550 behavior and dynamics as well as the communication interface
 551 with field-bus protocols. The power and energy optimization
 552 control algorithms are developed and downloaded to a B&R
 553 brand X20CP3586 programmable logic controller (PLC)-based

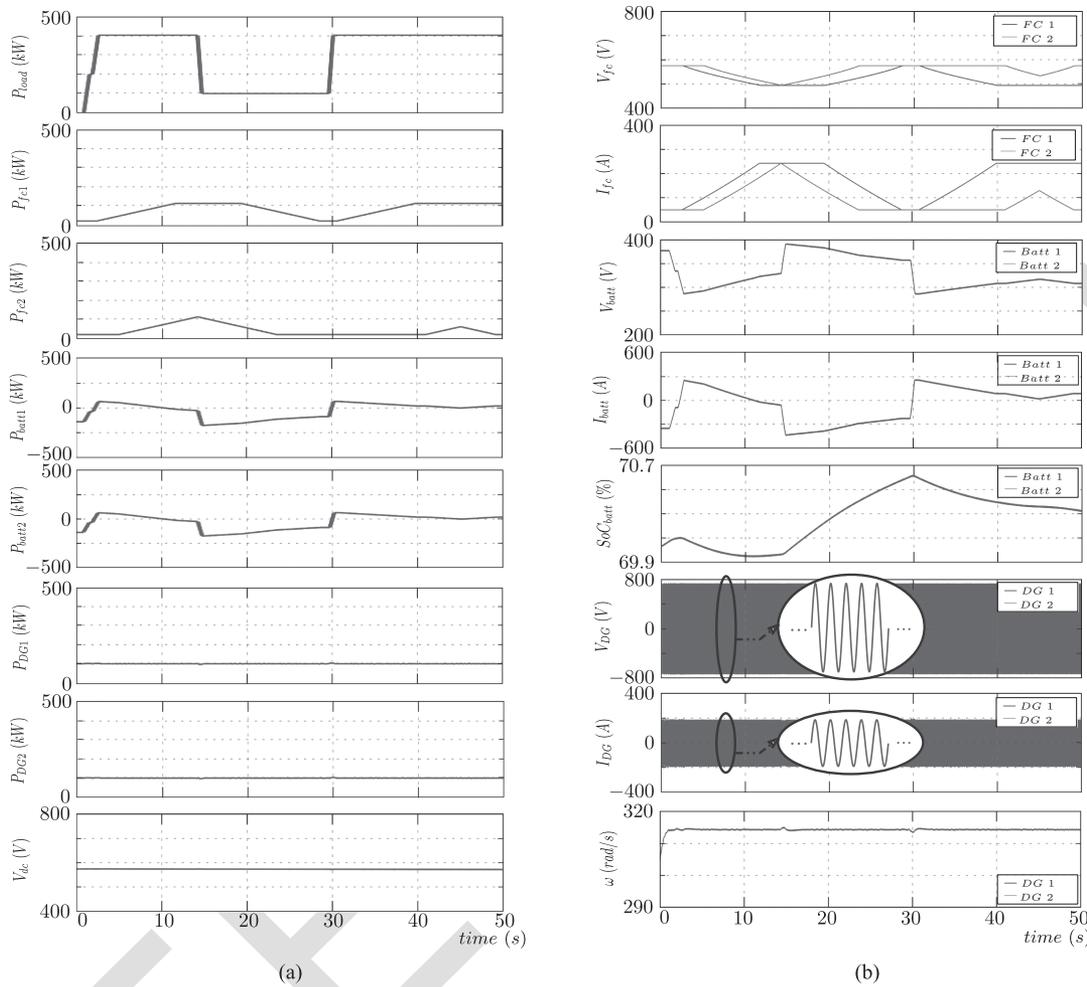


Fig. 12. Simulation results for the dc-based shipboard power grid model for demo vessel. DG in-parallel mode. Frequency of the DG is 50 Hz. (a) Power split of the hybrid shipboard power plant model. (b) Dynamic behavior of fuel cell devices, batteries and DGs.

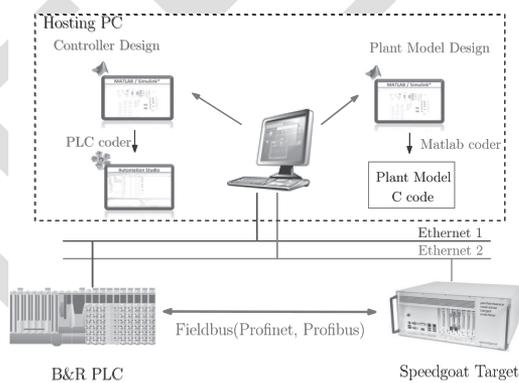


Fig. 13. HIL topology scheme.

554 embedded control system. The available real-time Fieldbus communication protocols between the plant model and controller are
 555 Profinet and Profibus for this setup. In this study, the topology
 556 diagram of the HIL test bed is shown in Fig. 13.
 557

558 V. SYSTEM MODEL VALIDATION

559 In this study, the mathematical shipboard power system model
 560 and HIL plant are both validated against the full-scale hybrid

power plant system. The actual power grid facilities are located
 561 at ABB collaborated hybrid power laboratory in MARINTEK,
 562 at Norwegian University of Science and Technology, in Trond-
 563 heim, Norway. The main objective of this experimental valida-
 564 tion test is to verify that the response of the modeled shipboard
 565 power system matches with the actual system response. The
 566 DGs and fuel cells with battery energy storage will be validated,
 567 respectively, according to the actual lab setup.
 568

The system configuration of the MARINTEK laboratory is
 569 based on a full-scale hybrid AES with ABB onboard dc-grid
 570 system. The lab arrangement diagram is shown in Fig. 14. The
 571 lab is equipped with two 410-kW diesel engines and variable
 572 speed generator sets, two 30-kW fuel cell devices, plus one
 573 55-kWh battery ESS. Power sources are connected to the dc-bus,
 574 and system loading conditions are simulated with two control-
 575 lable electric motors. Fig. 15 shows the equipment setup in
 576 MARINTEK lab.
 577

In the first validation test, one DG and one 55-kWh battery
 578 system are in use. The validation test results are shown in Fig. 16.
 579 The load profile is designed as four ramp-up cycles. The dynamic
 580 response of a DG is limited by its mechanical characteristic,
 581 and so the battery banks are there to boost the system dynamic
 582 performance to achieve the load demands. In this test, the
 583

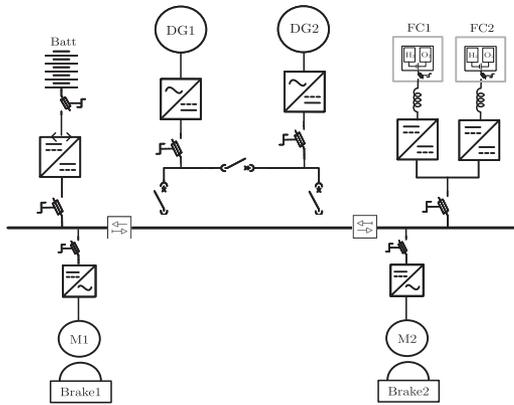


Fig. 14. ABB MARINTEK lab setup diagram, where M1 & M2 are the propulsion motors, and Brake1 & 2 present the controllable thrusters load.

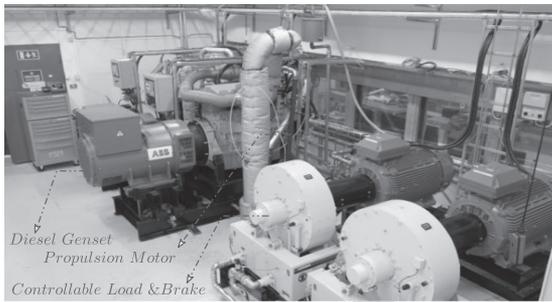


Fig. 15. ABB MARINTEK lab in Trondheim, Norway.

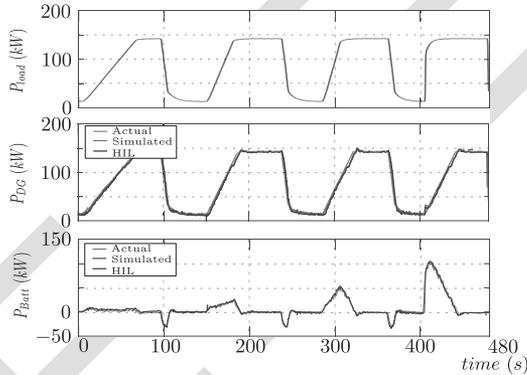


Fig. 16. Validation test I: To compare the system response between simulation model and HIL plant target in DG in-parallel mode.

584 simulated results from the Simulink model can match well with
 585 the actual recorded test. The overall mean average percentage
 586 error between the mathematical plant model and actual system
 587 is around 0.51%, and the average error between HIL plant and
 588 actual system is about 1.82%.

589 In the second validation test, two fuel cell modules and the
 590 same battery system are configured. The test results are shown in
 591 Fig. 17. During the first 100 s, the two fuel cells were powered up
 592 with 9-kW output one after another. The power drawn from the
 593 batteries was reduced with the increasing power supply from
 594 the fuel cells. From the 120 s, the load demands are reduced
 595 below the fuel cell power supply so that the batteries are charged
 596 by the fuel cells. Between 100 to 480 s, the fuel cells provide
 597 the constant power output, while the batteries absorb the load

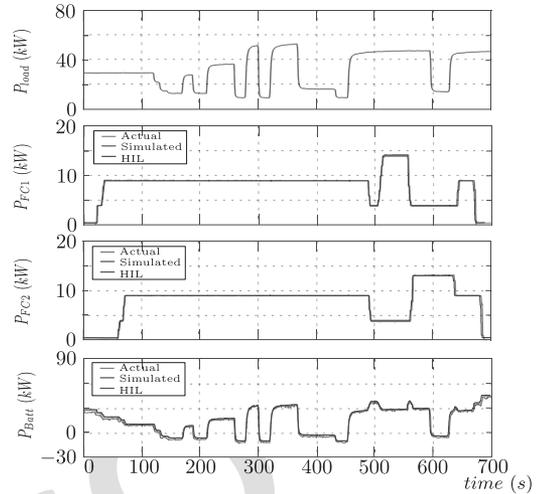


Fig. 17. Validation test II: To compare the system response between simulation model and HIL plant target in Zero-emission mode.

598 variations. From the 480 s onward, the power references for fuel
 599 cells are altered so that the batteries provide the remaining power
 600 to the load. The mean average percentage error is about 1.98%
 601 between the mathematical model and the actual system, while it
 602 is about 1.47% between HIL plant and the actual system.

603 To compare the test results, it can be seen that the system
 604 simulation and HIL results compare well with the experimental
 605 values recorded from the site. In addition, the transient rise time
 606 and steady-state power output for all DG, fuel cells, and battery
 607 power sources for both mathematical model and HIL plant can be
 608 in line with the actual system responses. Therefore, the proposed
 609 hybrid shipboard power plant model can be used to represent the
 610 physical system behavior.

VI. CONCLUSION

611 In this article, ad-c-distributed hybrid shipboard power system
 612 configuration has been studied. A system-level fuel cell-fed
 613 hybrid shipboard power grid system with dc distribution has
 614 been modeled. In addition, an HIL plant model platform has
 615 been set up and tested in a lab environment. Finally, both
 616 the system mathematical model and the HIL plant have been
 617 validated against a full-scale hybrid shipboard power system.
 618 This research work provides an essential platform and solid
 619 foundation to apply optimization control of power and energy
 620 management strategy in future work.

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